

Robust control of the boost-pressure of a turbocharger subject to model uncertainty and disturbances

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Abstract—Controller design for the boost-pressure of an exhaust-gas turbo-charger is considered. To this end, we propose a sliding mode controller with integral action in the sliding-variable. The design is simple, requires a small number of sensors and is robust with respect to various uncertainties and disturbances. Compared to standard integral sliding-mode controllers our design approach yields a dynamics of lower order which simplifies the stability and invariance analysis.

I. INTRODUCTION

The use of turbochargers increases efficiency and reduces emissions of modern combustion engines. Turbocharged combustion engines are widely applied in a broad field of applications. This includes building and agricultural machinery, locomotive and ship engines and also modern power systems, like block thermal power plants or gas turbine systems and, of course, automotive systems. This may be the reason for the sustained interest in research on the control of turbochargers, see [1], [2], [3], [4], [5], [6], [7] to name but a few.

A schematic for the airpath of a typical combustion engine is shown in Fig. 1. The exhaust-gas of the combustion engine on the right drives a turbine which is connected with a compressor on the fresh air side of the airpath. The compressor increases the air pressure in the intake manifold and thus enhances its capability of carrying oxygen to the combustion. In order to manipulate the power of the compressor, the turbine is equipped with a so-called variable-turbine geometry (VTG) which allows to vary the effective area of the turbine. The control task considered in this contribution is to control the boost-pressure in the intake manifold using the VTG as actuator.

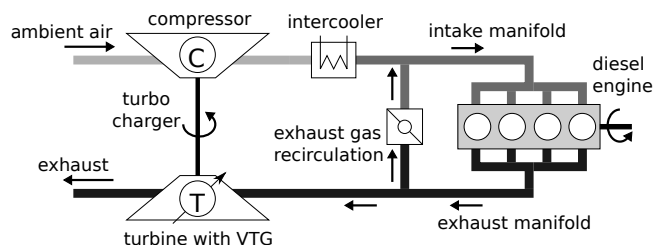


Fig. 1. Schematic of the airpath with turbocharger

Physical modeling and parameter-identification of a turbo-charger system is a difficult and time-consuming task

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[8],[9],[10]. Due to its thermodynamics nature the process exhibits strong nonlinearities. Furthermore, the system parameters also depend on the construction of the airpath and the combustion engine. Therefore, the parameter identification of such systems typically requires large experimental effort and is prone to significant uncertainties. Furthermore, process information at runtime is often restricted to a few sensor signals.

In this contribution we consider the control of the boost-pressure subject to model uncertainties and disturbances. In particular we shall consider constant parameter uncertainties and time-varying input disturbances. In our case-study, only measurements of the boost-pressure p_i and the exhaust-gas pressure p_e are available for the control.

Uncertainties and disturbances are called *matched* if they are within the input-span [11], otherwise they are called *unmatched*. Most parameter uncertainties in the airpath cause both types of uncertainties. It is well known that sliding-mode controllers may be successful in compensating (bounded) matched uncertainties. Unmatched uncertainties, in general, cannot be compensated by simple sliding-mode techniques [12]. However, with integral sliding-mode its influence can be minimised [11],[13],[14].

One of the first studies on the boost-pressure control using sliding-mode techniques is published in [15]. The authors devise a first-order sliding-mode controller combined with a state-observer. Attractivity of the sliding-manifold is guaranteed by the reaching law, however, there is no formal proof of stability of the dynamics on the sliding-manifold. In [16], [17] a MIMO approach is considered. Two control inputs, namely the exhaust-gas recirculation and VTG, are used to control the mass flow in the input-manifold as well as the exhaust-gas pressure as proposed in [18]. To the best knowledge of the authors, a study dedicated to the impact of various uncertainties for the boost-pressure control is not available to date.

Our approach is to use the VTG-actuator only and build our control design on the measurements of the boost-pressure and exhaust-gas pressure. We choose a sliding-manifold with integral action similar to the PID sliding-surface considered in [19] with some important modifications. The reduced dynamics of our approach are of third-order which allows to handle the stability analysis more easily. Classical integral sliding-mode techniques [13], [14] with an integrator in the nominal control yield a fourth-order dynamics on the sliding manifold, rendering the stability problem more complex.

The paper is organized as follows. In the subsequent section we recall the well-known third-order model of the

airpath that shall be subject of our investigations. In Section III we propose our controller. We describe the system in regular form and choose a sliding-manifold with integral action on p_i . The reduced dynamics on the sliding-manifold is of third order. Stability of the closed-loop system is guaranteed by a quadratic Lyapunov function. In Section IV we highlight the relation to similar sliding-mode techniques like PID-sliding manifold design proposed in [19] or integral sliding-mode in [13], [14]. Section V presents simulation results of the proposed controller in comparison to two other control approaches. We investigate the control performance for matched and unmatched uncertainties. Conclusions are drawn in Section VI.

II. MODEL OF THE AIRPATH

Subject of our analysis is the well-established model for the airpath in [20], [18] with parameters given in [4]. The third-order model is derived by considering the mass-flow balances in the respective volumes, expressed by

$$\dot{p}_i = \frac{RT_i}{V_i} (W_{ci} + W_{xi} - W_{ie}) \quad (1)$$

$$\dot{p}_x = \frac{RT_x}{V_x} (W_{ie} + W_f - W_{xi} - W_{xt}) \quad (2)$$

$$\dot{P}_c = \frac{1}{\tau} (-P_c + \eta_m P_t) \quad (3)$$

where p_i denotes the boost-pressure in the input manifold, p_x the pressure in the output manifold and P_c the power of the compressor. W_{ci} is the flow rate from the compressor to the input manifold. W_{xi} is the flow rate from the output manifold to the input manifold and describes the exhaust gas recirculation. W_{ie} is the flow rate from the input manifold into the engine and W_{xt} is the flow rate from the output manifold to the turbine.

The flow rates and the turbine power P_t can be determined as in [21]. They are given by

$$W_{ci} = \frac{\eta_c}{c_p T_a} \frac{P_c}{\left(\frac{p_i}{p_x}\right)^\mu - 1}, \quad (4)$$

$$W_{xi} = \frac{A_{egr} (u_{egr}) p_x}{\sqrt{RT_x}} \sqrt{\frac{2p_i}{p_x} \left(1 - \frac{p_i}{p_x}\right)}, \quad (5)$$

$$W_{ie} = \eta_v \frac{p_i n_m V_d}{2RT_i}, \quad (6)$$

$$W_{xt} = (c_1 u_{vtg} + c_2) \left(c_3 \left(\frac{p_a}{p_x} - 1 \right) + c_4 \right) \frac{p_x}{p_r} \sqrt{\frac{T_r}{T_x}} \sqrt{\frac{2p_i}{p_x} \left(1 - \frac{p_a}{p_x}\right)}, \quad (7)$$

$$P_t = W_{xt} c_p T_x \eta_t \left(1 - \left(\frac{p_a}{p_x}\right)^\mu\right). \quad (8)$$

The actuating variable is the VTG-position u_{vtg} which influences the turbine flow rate W_{xt} . In this paper, we shall consider the nominal case where no exhaust gas recirculation is considered, i.e. $W_{xi} = 0$. The systems states are required to remain in an admissible region given by:

$$p_i \in [p_a, 3 \cdot 10^5 \text{ Pa}], p_x \in [p_a, 6 \cdot 10^5 \text{ Pa}], P_c \in [0, 15 \text{ kW}]. \quad (9)$$

TABLE I
MODEL PARAMETERS FROM [4]

	Description	Value	Unit
n_m	Engine speed	41.66	[s ⁻¹]
W_f	Mass-flow fuel	2.5e-3	[kgs ⁻¹]
W_{ci}	Mass-flow compressor	-	[kgs ⁻¹]
W_{xi}	Mass-flow EGR	-	[kgs ⁻¹]
W_{ie}	Mass-flow intake manifold	-	[kgs ⁻¹]
W_{xt}	Fuel rate exhaust manifold	-	[kgs ⁻¹]
R	Specific gas constant	287	[Jkg ⁻¹ K ⁻¹]
p_i	Intake manifold pressure	-	[Pa]
p_x	Exhaust manifold pressure	-	[Pa]
p_r	Reference pressure	1.013e5	[Pa]
p_a	Ambient pressure	1.013e5	[Pa]
P_c	Compressor power	-	[kW]
P_t	Turbine power	-	[kW]
T_i	Intake manifold temperature	313	[K]
T_x	Exhaust manifold temperature	509	[K]
T_r	Reference temperature	298	[K]
T_a	Ambient temperature	298	[K]
V_i	Intake manifold volume	0.006	[m ³]
V_x	Exhaust manifold volume	0.001	[m ³]
V_d	Displacement volume	0.002	[m ³]
η_m	Turbo mechanical efficiency	98	[%]
η_c	Compressor isentropic efficiency	61	[%]
η_t	Turbine isentropic efficiency	76	[%]
η_v	Volumetric efficiency	87	[%]
c_p	Specific heat at constant pressure	1014.4	[Jkg ⁻¹ K ⁻¹]
μ	Constant	0.286	[-]
τ	Time-constant of turbocharger	0.11	[s]
c_1	VTG Parameter	-0.136	[-]
c_2	VTG Parameter	0.176	[-]
c_3	VTG Parameter	0.4	[-]
c_4	VTG Parameter	0.6	[-]

The complete set of parameters is listed in Table I.

III. CONTROL DESIGN

For the control design we consider the system in regular form and choose a sliding-surface with integral action that will compensate stationary effects of unmatched disturbances. Suitable initialisation of the integrator state allows to eliminate the reaching phase such that the system is in the sliding phase at all times. The resulting reduced dynamics on the sliding-manifold is of third order. Stability of the reduced dynamics is guaranteed by a quadratic Lyapunov function.

A. Regular form and disturbances

Neglecting the actuator dynamics we shall consider the exhaust-gas mass flow W_{xt} as input variable denoted by $u = W_{xt}$ in the remainder of this contribution. For the design of a sliding-mode controller it is convenient to consider the system in the so-called regular form proposed in [22] and [23]. Following the assessment in [24] we introduce the new state

$$z := P_c + \frac{\delta}{\zeta} \int_{p_a}^{p_x} \Omega(p) dp \quad (10)$$

where

$$\delta := \frac{\eta_m c_p T_x \eta_t}{\tau}, \quad \zeta := \frac{RT_x}{V_x}, \quad \Omega(p_x) := 1 - \left(\frac{p_a}{p_x}\right)^\mu.$$

With further abbreviations

$$\alpha := \frac{V_d \eta_i}{2V_i}, \quad \beta := \frac{RT_i \eta_c}{V_i c_p T_a}, \quad \gamma := \frac{\eta_V V_d T_x}{2V_x T_i}$$

$$\Phi(p_i) := \frac{\beta}{\left(\frac{p_i}{p_a}\right)^\mu - 1}, \quad \Psi(p_x) := \frac{\delta}{\zeta} \int_{p_a}^{p_x} \Omega(p) dp,$$

we obtain the system in regular form:

$$\dot{p}_i = -\alpha n_m p_i + \Phi(p_i)(z - \Psi(p_x)) \quad (11)$$

$$\dot{z} = \frac{\delta}{\zeta} \Omega(p_x)(\gamma n_m p_i + \zeta W_f) - \frac{1}{\tau}(z - \Psi(p_x)) \quad (12)$$

$$\dot{p}_x = \gamma n_m p_i + \zeta W_f - \zeta u. \quad (13)$$

Note that $\Omega(p_x) \in [0, 1]$ and Ω and Ψ are monotonically increasing functions, while Φ is monotonically decreasing and unbounded for $p_i \rightarrow p_a$. For each setpoint p_i^* of the system (11)-(13) there exists a unique equilibrium (p_i^*, z^*, p_x^*) and corresponding stationary control u^* . The admissible region (9) transformed into the new coordinates (p_i, z, p_x) shall be denoted by \mathbb{X} .

For the turbo-charger system in regular form (11)-(13), unmatched uncertainties enter the system in equation (11) or (12), whereas the matched disturbance adds to the right-hand side of (13), that is

$$\dot{p}_i = -\alpha n_m p_i + \Phi(p_i)(z - \Psi(p_x)) + \phi_{u1} \quad (14)$$

$$\dot{z} = \frac{\delta}{\zeta} \Omega(p_x)(\gamma n_m p_i + \zeta W_f) - \frac{1}{\tau}(z - \Psi(p_x)) + \phi_{u2} \quad (15)$$

$$\dot{p}_x = \gamma n_m p_i + \zeta W_f - \zeta u + \phi_m, \quad (16)$$

where ϕ_m denotes the matched and ϕ_{u1}, ϕ_{u2} the unmatched disturbance. The unmatched disturbance is state-dependent, the matched disturbance may be state and/or time-dependent.

In a practical application, we expect to face both types of disturbances. For instance, a parameter uncertainty of c_2 in (7) will result in some matched disturbance ϕ_m , whereas inspection of the system in regular form (11)-(13) reveals that most other parameter uncertainties lead to an unmatched uncertainty ϕ_{u1} or ϕ_{u2} or a combination of both. One of our major concerns are uncertainties that do not vanish at the nominal equilibrium point. Thus, the equilibrium may change due to the uncertainty. These kind of uncertainties may prohibit that the boost-pressure of the closed loop system converges to the set-point. We propose a control law that drives the system to its setpoint despite the presence of such uncertainties and also compensates the matched (possibly time-varying) disturbance.

B. Proposed sliding-mode controller

Sliding mode control ensures robustness to all bounded matched uncertainties. In order to address unmatched uncertainties which affect the stationary value of the boost-pressure, an additional integrator state can be introduced and used in the control law.

The idea of choosing a PID-type sliding-surface goes back to [25] and [19] and has been extended and applied in various

applications, see e.g. [26], [27] and [28]. Adopted for our system such sliding-surface takes the form:

$$s = k_p(p_i - p_i^*) + k_i v + k_d \frac{d}{dt}(p_i - p_i^*)$$

$$= k_p(p_i - p_i^*) + k_i v + k_d(-\alpha n_m p_i + \Phi(p_i)(z - \Psi(p_x)))$$

with $\dot{v} = p_i - p_i^*$. The control law is typically obtained by requiring $\dot{s} = 0$. Note, that the derivative-part k_d in s is essential in order to obtain the control input u in \dot{s} .

Inspection of the equations above reveals two drawbacks of this approach for our application scenario: Firstly, the control law derived from the second equation will contain the state z and thus the measurement P_c , c.f. (10), which is not available; secondly, we need to take the derivative of the signal p_i . This can cause additional problems in the presence of measurement noise.

Therefore we choose a much simpler sliding-surface and incorporate a second state p_x :

$$s = p_x - p_x^* - v \quad (17)$$

$$\dot{v} = -k(p_i - p_i^*), \quad v(0) = v_0, \quad (18)$$

with $k > 0$. Note that additional design parameters in s do not increase the freedom in shaping the reduced dynamics. Consequently, we choose the control law as the superposition of the equivalent control and a discontinuous control [12]:

$$u = \frac{1}{\zeta} \left(\gamma n_m p_i + \zeta W_f + k(p_i - p_i^*) + \rho \operatorname{sgn}(s) \right). \quad (19)$$

The choice of $\rho > |\phi_m|$ for all times guarantees the attractivity of the sliding manifold $s = 0$.

On the sliding-manifold $s = 0$ holds $v = p_x - p_x^*$ and thus $\dot{p}_x = -k(p_x - p_x^*)$. From (11)-(13) we obtain the reduced dynamics

$$\dot{p}_i = -\alpha n_m p_i + \Phi(p_i)(z - \Psi(p_x)) \quad (20)$$

$$\dot{z} = \frac{\delta}{\zeta} \Omega(p_x)(\gamma n_m p_i + \zeta W_f) - \frac{1}{\tau}(z - \Psi(p_x)) \quad (21)$$

$$\dot{p}_x = -k(p_x - p_x^*). \quad (22)$$

If the sliding-manifold is attractive and the reduced dynamics is asymptotically stable, at the equilibrium we have

$$\dot{p}_x = 0 \Rightarrow p_x = p_x^*.$$

Note that this last result is robust against any unmatched disturbance ϕ_{u1}, ϕ_{u2} . Matched disturbances ϕ_m are fully compensated if $\rho > |\phi_m|$.

C. Stability analysis

By choosing $v_0 = p_x(0) - p_x^*(0)$ we can ensure that the trajectory starts on the sliding manifold. Thus, asymptotic stability of the reduced dynamics (20)-(22) guarantees stability of the closed-loop system (11)-(13) and (19).

Consider the nominal case, i.e. $\phi_{u1} = \phi_{u2} = \phi_m = 0$. We show stability and determine a part of the region of attraction using a quadratic Lyapunov function that we draw

from the linearised system. Linearisation at the equilibrium (p_i^*, z^*, p_x^*) yields the dynamics matrix

$$A = \begin{bmatrix} -\alpha n_m + \frac{d}{dp_i} \Phi(p_i^*)(z^* - \Psi(p_x^*)) & \Phi(p_i^*) \\ \frac{\delta}{\zeta} \Omega(p_x^*) \gamma n_m & -\frac{1}{\tau} \\ -k & 0 \\ -\Phi(p_i^*) \frac{\delta}{\zeta} \Omega(p_x^*) & \\ \frac{\delta}{\zeta} \frac{d}{dp_x} \Omega(p_x^*) (\gamma n_m p_i^* + \zeta W_f) + \frac{\delta}{\tau \zeta} \Omega(p_x^*) & \\ 0 & \end{bmatrix}.$$

An estimate of the region of attraction for the nonlinear system is obtained by the largest admissible level set for which $x^T P f(x) < 0$, where $x = (p_i, z, p_x)^T$. f denotes the right-hand side of (20)-(22), and P is symmetric and positive definite solution of the corresponding Lyapunov equation.

It turns out that the level-sets tend to span only a small interval in the p_x -direction. Therefore we formulate the following optimisation problem:

$$\min_{q_2, q_3 > 0} -p_{x \text{opt}}(q_2, q_3)$$

subject to

$$A^T P + P A = - \begin{pmatrix} 1 & 0 & 0 \\ 0 & q_2 & 0 \\ 0 & 0 & q_3 \end{pmatrix},$$

$$\Omega_c = \{x \in \mathbb{R}^3 \mid x^T P x = c\},$$

$$c_{max} = \max \{c \in \mathbb{R}^+ \mid \forall x \in \Omega_c \cap \mathbb{X}: x^T P f(x) < 0\},$$

$$p_{x \text{opt}}(q_2, q_3) = \max \{p_x \in \mathbb{R}^+ \mid p_x \in \Omega_{c_{max}}\}.$$

For the setpoint $p_i^* = 2000$ the determined region of attraction is shown in Fig. 2. The outlined shape depicts the admissible region of safe operation of the turbo-charger. While the determined level-set is not a strict subset of the admissible region \mathbb{X} , we can verify numerically via a simple invariance condition that the intersection of the admissible region and this level set is positively invariant. This guarantees that the intersection is part of the region of attraction of the equilibrium point.

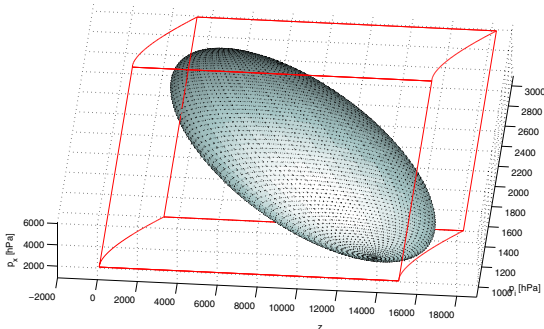


Fig. 2. Estimated region of attraction and invariant admissible region \mathbb{X}

IV. RELATION TO INTEGRAL SLIDING-MODE

The proposed control design belongs to the class of integral sliding-mode (ISM) controllers defined in [11]. In recent years a design procedure has been proposed that extends (any) nominal control law u_0 by a discontinuous control u_1 , minimising the impact of the unmatched uncertainty, [13], [29], [14]. In regard to our problem (11)-(13), the nominal control is to address the unmatched uncertainty whereas the integral sliding-mode controller eliminates the matched uncertainty. In the following we shall design such a ISM control and cast our proposed controller into this framework.

A. Classical design

In order to compensate time-invariant unmatched uncertainties, we may choose a PI-state-feedback controller for the nominal control u_0 , that is

$$\dot{v} = p_i - p_i^* \quad (23)$$

$$u_0 = k_{sf}^T (p_i \quad z \quad p_x)^T + k_i v, \quad (24)$$

where $k_{sf} \in \mathbb{R}^3$. Following [13], [14], optimal disturbance rejection is achieved choosing the sliding manifold:

$$s = p_x - p_x^* + v_2.$$

This leads to integrator state and discontinuous dynamics

$$\begin{aligned} \dot{v}_2 &= -(\gamma n_m p_i + \zeta W_f - \zeta u_0), \\ u_1 &= \frac{\rho}{\zeta} \text{sgn}(s). \end{aligned}$$

On the sliding surface we obtain the reduced dynamics

$$\dot{p}_i = -\alpha n_m p_i + \Phi(p_i)(z - \Psi(p_x)) \quad (25)$$

$$\dot{z} = \frac{\delta}{\zeta} \Omega(p_x) (\gamma n_m p_i + \zeta W_f) - \frac{1}{\tau} (z - \Psi(p_x)) \quad (26)$$

$$\dot{p}_x = \gamma n_m p_i + \zeta W_f + k_{sf}^T (p_i \quad z \quad p_x)^T + k_i v \quad (27)$$

$$\dot{v} = p_i - p_i^*. \quad (28)$$

When constrained to the sliding-surface, the system behaves just like the nominal closed-loop system with PI-state-feedback control. Assuming a suitable design, the matched uncertainty has no impact and the unmatched time-invariant uncertainty does not influence the stationary behaviour of the input pressure. However, the reduced dynamics is of fourth-order as compared to the third-order dynamics (20)-(22) of our proposed approach. This renders the stability analysis much more complex.

B. Proposed design in form of classical ISM

The resulting dynamics of the classical ISM design is completely defined by the ideal control u_0 and the choice of the switching function in the presence of unmatched uncertainties. Thus, ISM can be seen as an addition to an existing control which ensures robustness to matched uncertainties. With regard to unmatched uncertainties, this means that the ISM control does not exceed the performance of the control u_0 alone.

We have chosen a different approach to design the ISM controller. Instead of minimising the effect of the unmatched

uncertainty in comparison to the ideal control, we use the switching function to shape the reduced dynamics. The integrator state v is used to ensure the stationary accuracy of the boost-pressure which, however, reduces the design freedom for the ideal control u_0 .

If we cast our approach into the classical design we obtain:

$$\begin{aligned} u_0 &= \frac{1}{\zeta} \left(\gamma n_m p_i + \zeta W_f + k(p_i - p_i^*) \right), \\ s &= p_x - p_x^* + v, \\ u_1 &= \frac{\rho}{\zeta} \operatorname{sgn}(s), \\ \dot{v} &= k(p_i - p_i^*). \end{aligned}$$

In light of this, our proposed control can be written in the sense of [14]. Our choice of the sliding manifold also minimises the effect of unmatched disturbances as proven in [13]. At a first glance, it might seem remarkable that u_0 does not contain integral action. Substituting u_0 into (13) reveals that p_x essentially serves this duty. A different interpretation may be that the integrator v of the sliding surface doubles up as integrator for the nominal control.

V. SIMULATION RESULTS

Simulation results shall illustrate the performance of our controller. We consider a time-invariant unmatched uncertainty and also a time-varying matched uncertainty. In order to highlight the robustness properties, we also design a simple PI state-feedback controller (capable of compensating constant model uncertainties) and a first-order sliding-mode controller (capable of compensating matched uncertainties). The PI state-feedback of the form (24) is devised such that the eigenvalues of linearised closed-loop dynamics are $\{-21, -21, -2, -2\}$. The sliding-surface and control law of the first-order SMC is obtained from (17), (18) and (19) with $k = 0, \rho = 2 \times 10^6$. In order to reduce chattering, we approximate the discontinuity $\operatorname{sgn}(s) \approx \frac{s}{|s|+1000}$. For our proposed controller we choose $k = 7, \rho = 2 \times 10^6$.

Note that the three controllers are not tuned for best performance, but to give similar rise and settling times to allow for a better comparison.

If proved to be stable, using the classical ISM discussed in Section IV-A we would expect a similar performance compared to our proposed approach.

We simulate the system with the proposed controllers under the influence of different parameter uncertainties and evaluate the results. In the first scenario, we consider the parameter c_2 to be uncertain by 10% and time-varying so that a matched disturbance affects the system (14)-(16) with:

$$\phi_m = -\zeta \tilde{c}_2 \sin(4\pi t) \left(c_3 \left(\frac{p_a}{p_x} - 1 \right) + c_4 \right), \quad (29)$$

with $\tilde{c}_2 = 0.1c_2$.

For the second scenario, we admit a parameter uncertainty of 10% in γ . With $\tilde{\gamma} = 0.1\gamma$ this causes unmatched and matched uncertainties given by:

$$\phi_{u1} = 0, \quad \phi_{u2} = \frac{\delta}{\zeta} \Omega(p_x) \tilde{\gamma} n_m p_i, \quad \phi_m = \tilde{\gamma} n_m p_i. \quad (30)$$

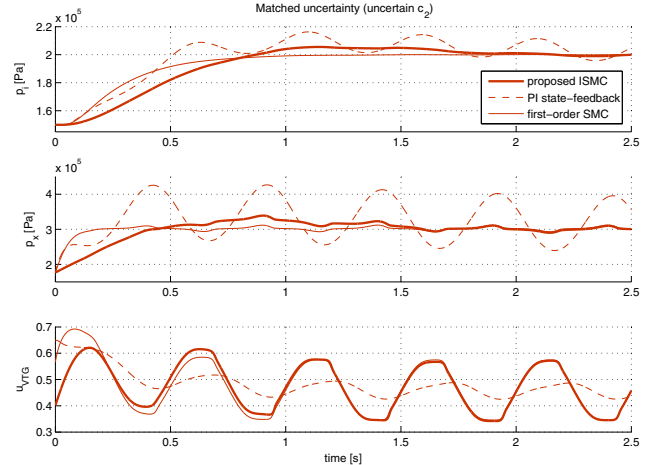


Fig. 3. Matched uncertainty entering via c_2

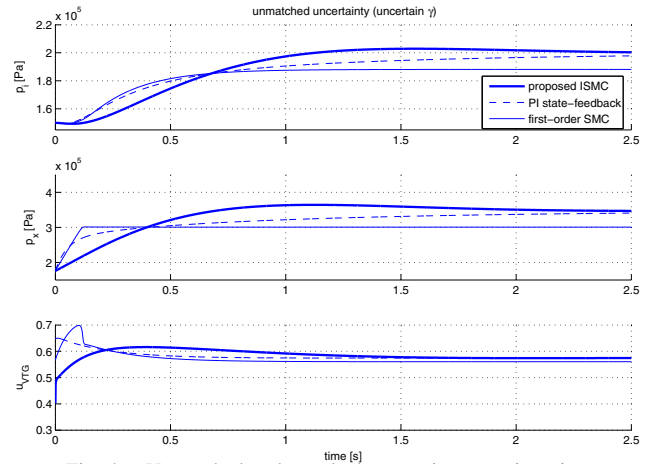


Fig. 4. Unmatched and matched uncertainty entering via γ

Fig. 3 shows the simulation results of the three controllers for the time-varying matched uncertainty ϕ_m in (29). The first two graphs show the states p_i and p_x , respectively. The third plot shows the normalised control signal. The system with PI state-feedback does not reach steady state due to persistent disturbance. Both sliding-mode controllers can compensate the matched uncertainty.

In Fig. 4 we see the results for the second scenario (30) with the uncertainty in γ . The PI state-feedback can compensate the stationary influence of both uncertainties on the input pressure. The first-order SMC, however, shows an offset in the boost-pressure p_i and the exhaust-gas pressure p_x . This drawback is not present for the proposed ISM controller.

Fig. 5 shows a comparison for each controller of the nominal undisturbed response and each of the two scenarios considered above. As expected, the two sliding-mode controllers show (almost) no differences to the nominal performance when a matched disturbance is added. However, our proposed ISM controller (2nd plot) shows the most consistent performance under uncertainties.

VI. CONCLUSION

We have devised a controller for the boost-pressure control in an exhaust-gas turbocharger system under uncertainties.

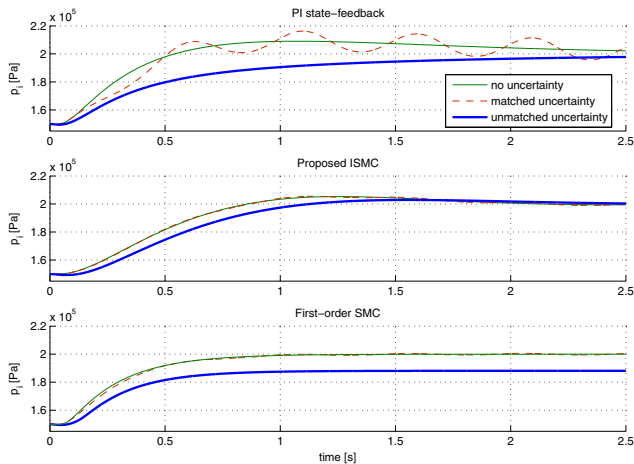


Fig. 5. Robustness of performance

Using the variable turbine geometry as the only control input, first, the system was transformed into regular form. Afterwards, the typical sliding mode design was augmented by an integrator to deal with the stationary effect of the unmatched uncertainty. We show stability of the controlled system using Lyapunov stability theory. In a comparison with classical ISM we clarified the different purposes of the integral state. Whereas in classical ISM the integrator state mainly allows an immediate start in sliding mode while retaining the performance of the nominal control, our design uses the integrator state to reduce the influence of the unmatched uncertainty. Still both design concepts can be reformulated into each other. Simulation results show that the proposed controller is robust with respect to matched and unmatched uncertainties. Further work may concentrate on the design of the sliding manifold, incorporating all states to improve the performance while retaining the robustness properties.

ACKNOWLEDGEMENT

Johann Reger kindly acknowledges support by the European Union Horizon 2020 research and innovation program under Marie Skłodowska-Curie grant agreement No. 734832.

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